

LM386 Low Voltage Audio Power Amplifier

General Description

The LM386 is a power amplifier designed for use in low voltage consumer applications. The gain is internally set to 20 to keep external part count low, but the addition of an external resistor and capacitor between pins 1 and 8 will increase the gain to any value up to 200.

The inputs are ground referenced while the output is automatically biased to one half the supply voltage. The quiescent power drain is only 24 milliwatts when operating from a 6 volt supply, making the LM386 ideal for battery operation.

Features

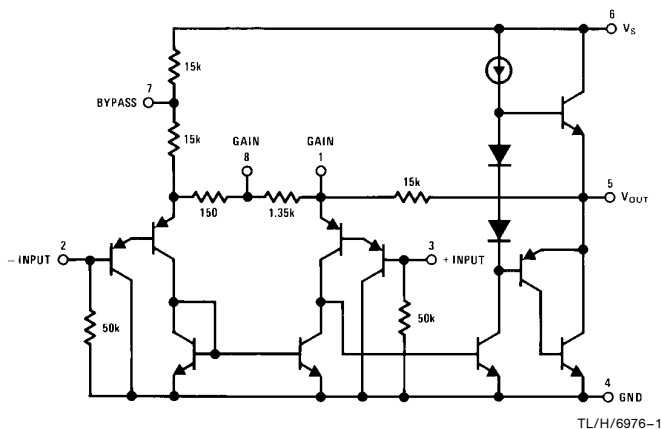
- Battery operation
- Minimum external parts
- Wide supply voltage range 4V–12V or 5V–18V
- Low quiescent current drain 4 mA

- Voltage gains from 20 to 200
- Ground referenced input
- Self-centering output quiescent voltage
- Low distortion
- Eight pin dual-in-line package

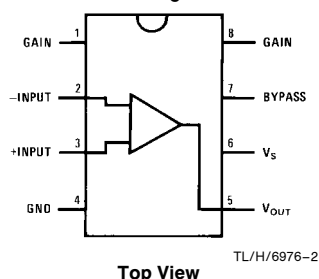
Applications

- AM-FM radio amplifiers
- Portable tape player amplifiers
- Intercoms
- TV sound systems
- Line drivers
- Ultrasonic drivers
- Small servo drivers
- Power converters

Equivalent Schematic and Connection Diagrams



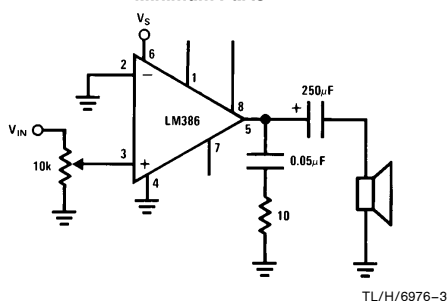
Dual-In-Line and Small Outline Packages



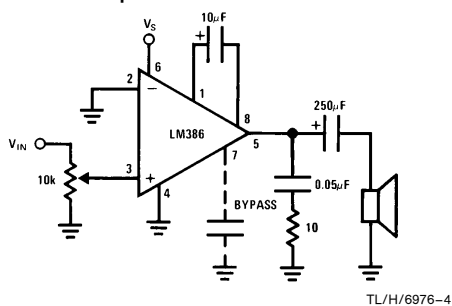
Order Number LM386M-1,
LM386N-1, LM386N-3 or LM386N-4
See NS Package Number
M08A or N08E

Typical Applications

Amplifier with Gain = 20 Minimum Parts



Amplifier with Gain = 200



Absolute Maximum Ratings

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage (LM386N-1, -3, LM386M-1)	15V
Supply Voltage (LM386N-4)	22V
Package Dissipation (Note 1) (LM386N)	1.25W
(LM386M)	0.73W
Input Voltage	$\pm 0.4V$
Storage Temperature	$-65^{\circ}C$ to $+150^{\circ}C$
Operating Temperature	$0^{\circ}C$ to $+70^{\circ}C$
Junction Temperature	$+150^{\circ}C$

Soldering Information

Dual-In-Line Package	
Soldering (10 sec)	$+260^{\circ}C$
Small Outline Package	
Vapor Phase (60 sec)	$+215^{\circ}C$
Infrared (15 sec)	$+220^{\circ}C$

See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering surface mount devices.

Thermal Resistance

θ_{JC} (DIP)	$37^{\circ}C/W$
θ_{JA} (DIP)	$107^{\circ}C/W$
θ_{JC} (SO Package)	$35^{\circ}C/W$
θ_{JA} (SO Package)	$172^{\circ}C/W$

Electrical Characteristics $T_A = 25^{\circ}C$

Parameter	Conditions	Min	Typ	Max	Units
Operating Supply Voltage (V_S) LM386N-1, -3, LM386M-1 LM386N-4		4 5		12 18	V V
Quiescent Current (I_Q)	$V_S = 6V, V_{IN} = 0$		4	8	mA
Output Power (P_{OUT}) LM386N-1, LM386M-1 LM386N-3 LM386N-4	$V_S = 6V, R_L = 8\Omega, THD = 10\%$ $V_S = 9V, R_L = 8\Omega, THD = 10\%$ $V_S = 16V, R_L = 32\Omega, THD = 10\%$	250 500 700	325 700 1000		mW mW mW
Voltage Gain (A_V)	$V_S = 6V, f = 1\text{ kHz}$ $10\ \mu F$ from Pin 1 to 8		26 46		dB dB
Bandwidth (BW)	$V_S = 6V$, Pins 1 and 8 Open		300		kHz
Total Harmonic Distortion (THD)	$V_S = 6V, R_L = 8\Omega, P_{OUT} = 125\text{ mW}$ $f = 1\text{ kHz}$, Pins 1 and 8 Open		0.2		%
Power Supply Rejection Ratio (PSRR)	$V_S = 6V, f = 1\text{ kHz}, C_{BYPASS} = 10\ \mu F$ Pins 1 and 8 Open, Referred to Output		50		dB
Input Resistance (R_{IN})			50		k Ω
Input Bias Current (I_{BIAS})	$V_S = 6V$, Pins 2 and 3 Open		250		nA

Note 1: For operation in ambient temperatures above $25^{\circ}C$, the device must be derated based on a $150^{\circ}C$ maximum junction temperature and 1) a thermal resistance of $80^{\circ}C/W$ junction to ambient for the dual-in-line package and 2) a thermal resistance of $170^{\circ}C/W$ for the small outline package.

Application Hints

GAIN CONTROL

To make the LM386 a more versatile amplifier, two pins (1 and 8) are provided for gain control. With pins 1 and 8 open the $1.35\text{ k}\Omega$ resistor sets the gain at 20 (26 dB). If a capacitor is put from pin 1 to 8, bypassing the $1.35\text{ k}\Omega$ resistor, the gain will go up to 200 (46 dB). If a resistor is placed in series with the capacitor, the gain can be set to any value from 20 to 200. Gain control can also be done by capacitively coupling a resistor (or FET) from pin 1 to ground.

Additional external components can be placed in parallel with the internal feedback resistors to tailor the gain and frequency response for individual applications. For example, we can compensate poor speaker bass response by frequency shaping the feedback path. This is done with a series RC from pin 1 to 5 (paralleling the internal $15\text{ k}\Omega$ resistor). For 6 dB effective bass boost: $R \approx 15\text{ k}\Omega$, the lowest value for good stable operation is $R = 10\text{ k}\Omega$ if pin 8 is open. If pins 1 and 8 are bypassed then R as low as $2\text{ k}\Omega$ can be used. This restriction is because the amplifier is only compensated for closed-loop gains greater than 9.

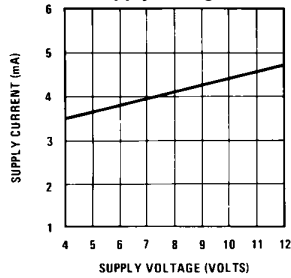
INPUT BIASING

The schematic shows that both inputs are biased to ground with a $50\text{ k}\Omega$ resistor. The base current of the input transistors is about 250 nA , so the inputs are at about 12.5 mV when left open. If the dc source resistance driving the LM386 is higher than $250\text{ k}\Omega$ it will contribute very little additional offset (about 2.5 mV at the input, 50 mV at the output). If the dc source resistance is less than $10\text{ k}\Omega$, then shorting the unused input to ground will keep the offset low (about 2.5 mV at the input, 50 mV at the output). For dc source resistances between these values we can eliminate excess offset by putting a resistor from the unused input to ground, equal in value to the dc source resistance. Of course all offset problems are eliminated if the input is capacitively coupled.

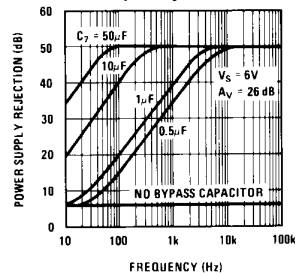
When using the LM386 with higher gains (bypassing the $1.35\text{ k}\Omega$ resistor between pins 1 and 8) it is necessary to bypass the unused input, preventing degradation of gain and possible instabilities. This is done with a $0.1\ \mu F$ capacitor or a short to ground depending on the dc source resistance on the driven input.

Typical Performance Characteristics

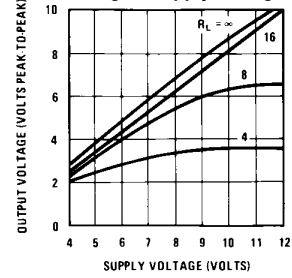
Quiescent Supply Current vs Supply Voltage



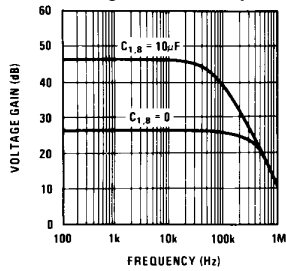
Power Supply Rejection Ratio (Referred to the Output) vs Frequency



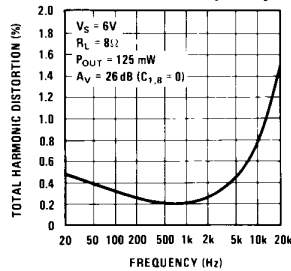
Peak-to-Peak Output Voltage Swing vs Supply Voltage



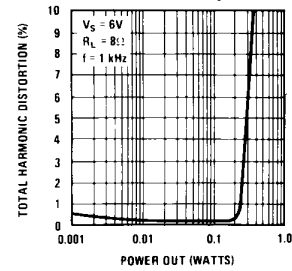
Voltage Gain vs Frequency



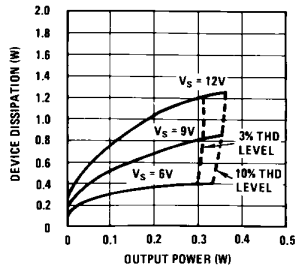
Distortion vs Frequency



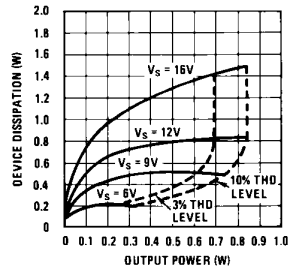
Distortion vs Output Power



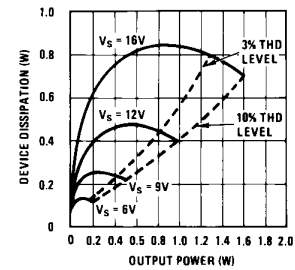
Device Dissipation vs Output Power—4 Ω Load



Device Dissipation vs Output Power—8 Ω Load



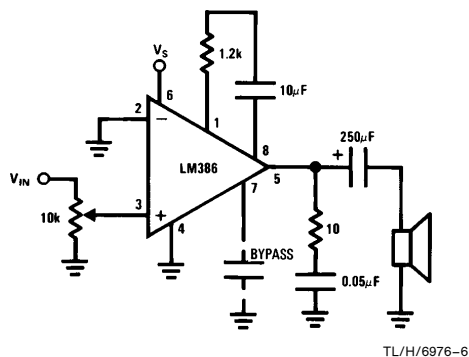
Device Dissipation vs Output Power—16 Ω Load



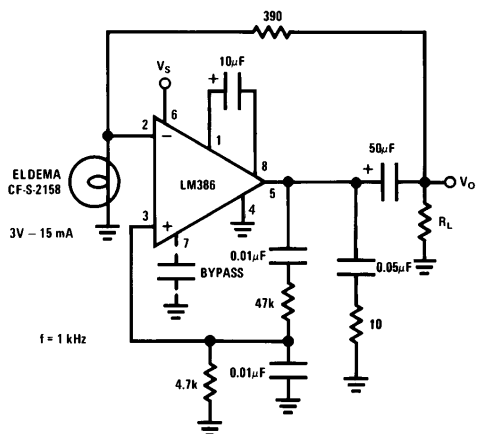
TL/H/6976-5

Typical Applications (Continued)

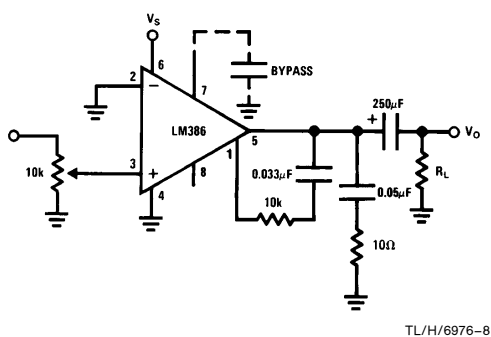
Amplifier with Gain = 50



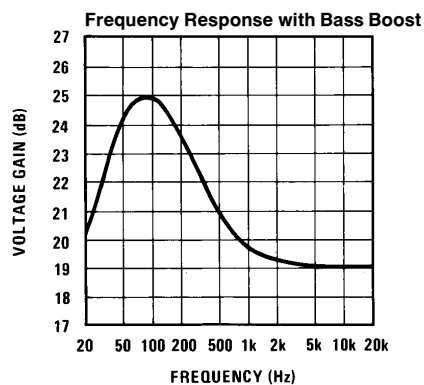
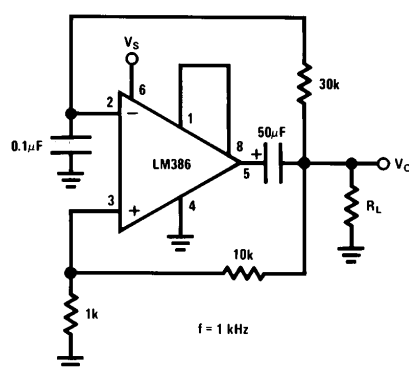
Low Distortion Power Wienbridge Oscillator



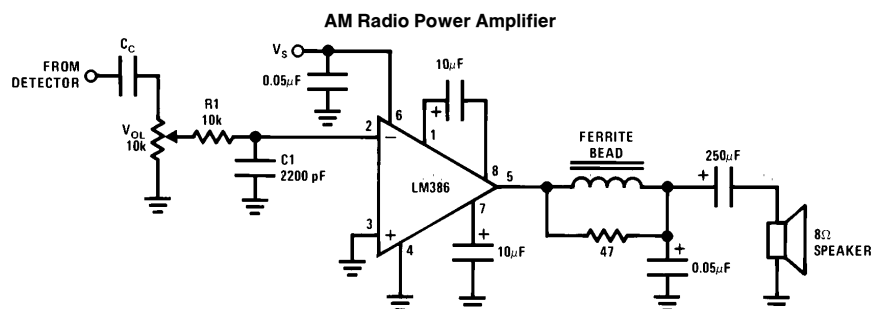
Amplifier with Bass Boost



Square Wave Oscillator



Typical Applications (Continued)



Note 1: Twist supply lead and supply ground very tightly.

Note 2: Twist speaker lead and ground very tightly.

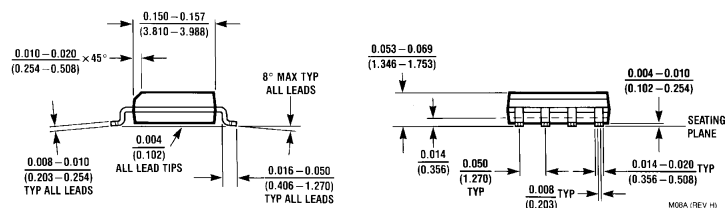
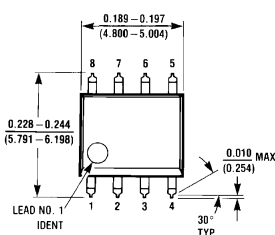
Note 3: Ferrite bead is Ferroxcube K5-001-001/3B with 3 turns of wire.

Note 4: R1C1 band limits input signals.

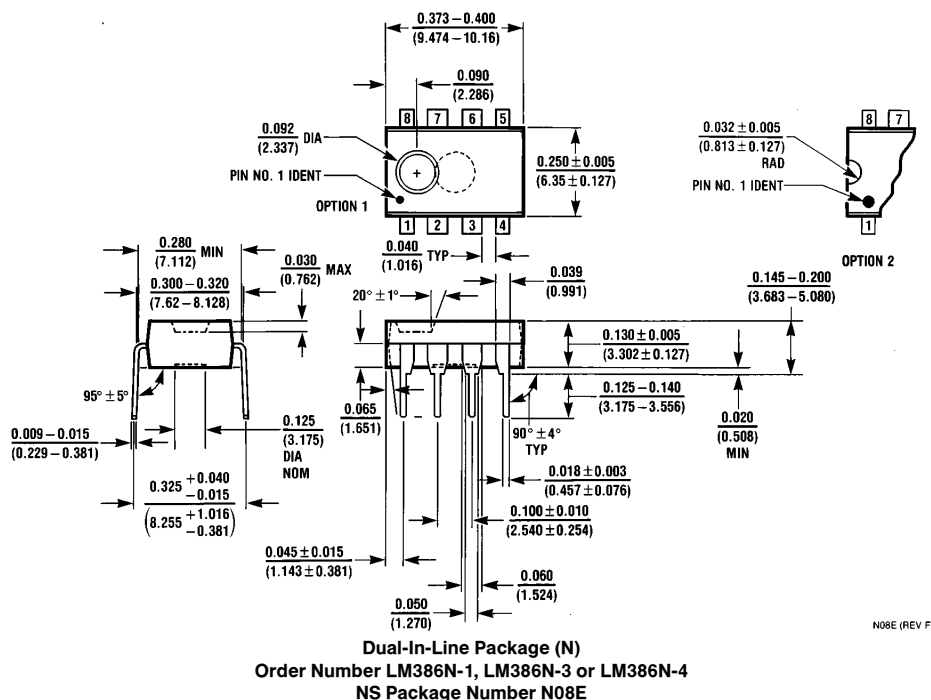
Note 5: All components must be spaced very close to IC.

TL/H/6976-11

Physical Dimensions inches (millimeters)



SO Package (M)
Order Number LM386M-1
NS Package Number M08A

Physical Dimensions inches (millimeters) (Continued)**LIFE SUPPORT POLICY**

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